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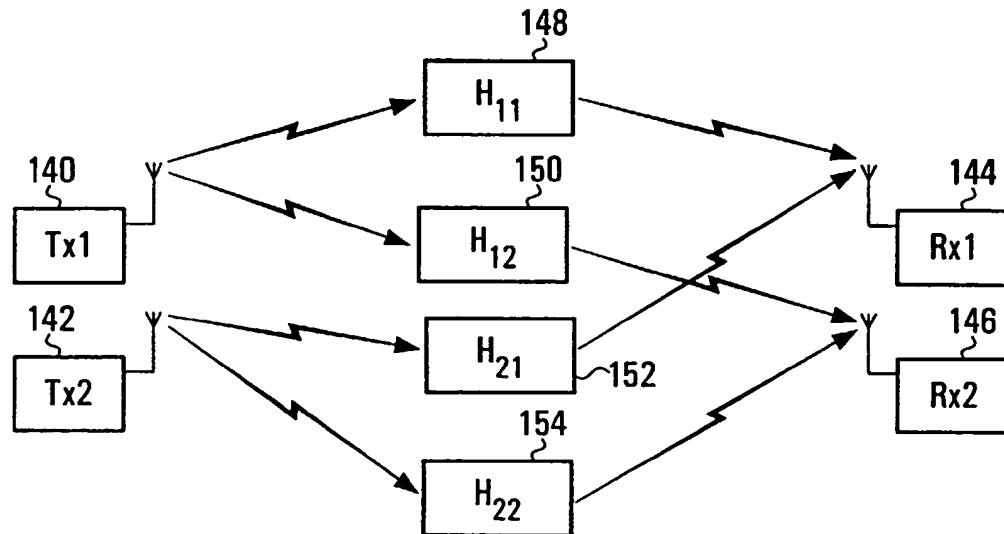
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(54) Title: SCATTERED PILOT PATTERN AND CHANNEL ESTIMATION METHOD FOR MIMO-OFDM SYSTEMS



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(57) Abstract: A method and apparatus are provided for reducing the number of pilot symbols within a MIMO-OFDM communication system, and for improving channel estimation within such a system. For each transmitting antenna in an OFDM transmitter, pilot symbols are encoded so as to be unique to the transmitting antenna. The encoded pilot symbols are then inserted into an OFDM frame to form a diamond lattice, the diamond lattices for the different transmitting antennae using the same frequencies but being offset from each other by a single symbol in the time domain. At the OFDM receiver, a channel response is estimated for a symbol central to each diamond of the diamond lattice using a two-dimensional interpolation. The estimated channel responses are smoothed in the frequency domain. The channel responses of remaining symbols are then estimated by interpolation in the frequency domain.



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SCATTERED PILOT PATTERN AND CHANNEL ESTIMATION METHOD FOR
MIMO-OFDM SYSTEMS

FIELD OF THE INVENTION

5 This invention relates to OFDM communication systems, and more particularly to a more efficient use of pilot symbols within such systems.

BACKGROUND OF THE INVENTION

Multiple Input Multiple Output - Orthogonal
10 Frequency Division Multiplexing (MIMO-OFDM) is a novel highly spectral efficient technology used to transmit high-speed data through radio channels with fast fading both in frequency and in time.

In wireless communication systems that employ
15 OFDM, a transmitter transmits data to a receiver using many sub-carriers in parallel. The frequencies of the sub-carriers are orthogonal. Transmitting the data in parallel allows the symbols containing the data to be of longer duration, which reduces the effects of multi-path fading.

20 The orthogonality of the frequencies allows the sub-carriers to be tightly spaced, while minimizing inter-carrier interference. At the transmitter, the data is encoded, interleaved, and modulated to form data symbols. Overhead information is added, including pilot symbols, and the
25 symbols (data plus overhead) are organized into OFDM symbols. Each OFDM symbol typically uses 2^n frequencies.

Each symbol is allocated to represent a component of a different orthogonal frequency. An inverse Fast Fourier Transform (IFFT) is applied to the OFDM symbol (hence the preference of 2^n frequencies) to generate time samples of a signal. Cyclic extensions are added to the signal, and the signal is passed through a digital-to-analog converter.

Finally, the transmitter transmits the signal to the receiver along a channel.

When the receiver receives the signal, the inverse operations are performed. The received signal is passed through an analog-to-digital converter, and timing information is then determined. The cyclic extensions are removed from the signal. The receiver performs an FFT on the received signal to recover the frequency components of the signal, that is, the data symbols. Error correction may be applied to the data symbols to compensate for variations in phase and amplitude caused during propagation of the signal along the channel. The data symbols are then demodulated, de-interleaved, and decoded, to yield the transmitted data.

In systems employing differential detection, the receiver compares the phase and/or amplitude of each received symbol with an adjacent symbol. The adjacent symbol may be adjacent in the time direction or in the frequency direction. The receiver recovers the transmitted data by measuring the change in phase and/or amplitude between a symbol and the adjacent symbol. If differential detection is used, channel compensation need not be applied to compensate for variations in phase and amplitude caused during propagation of the signal. However, in systems employing coherent detection the receiver must estimate the actual phase and amplitude of the channel response, and channel compensation must be applied.

The variations in phase and amplitude resulting from propagation along the channel are referred to as the channel response. The channel response is usually frequency and time dependent. If the receiver can determine the channel response, the received signal can be corrected to

compensate for the channel degradation. The determination of the channel response is called channel estimation. The inclusion of pilot symbols in each OFDM symbol allows the receiver to carry out channel estimation. The pilot symbols are transmitted with a value known to the receiver. When the receiver receives the OFDM symbol, the receiver compares the received value of the pilot symbols with the known transmitted value of the pilot symbols to estimate the channel response.

The pilot symbols are overhead, and should be as few in number as possible in order to maximize the transmission rate of data symbols. Since the channel response can vary with time and with frequency, the pilot symbols are scattered amongst the data symbols to provide as complete a range as possible of channel response over time and frequency. The set of frequencies and times at which pilot symbols are inserted is referred to as a pilot pattern. The optimal temporal spacing between the pilot symbols is usually dictated by the maximum anticipated Doppler frequency, and the optimal frequency spacing between the pilot symbols is usually dictated by the anticipated delay spread of multi-path fading.

The existing pilot-assisted OFDM channel estimation approaches are designed for conventional one transmitter system. With a scattered pilot arrangement, there are three classes of algorithms:

- 1-D frequency interpolation or time interpolation
- Transformed frequency 1-D interpolation
- Independent time and frequency 1-D interpolation

5 The first class of algorithms is based on the pilot OFDM symbol (all the sub-carriers are used as the pilots) or comb-type of pilots. This approach shown in the flow chart of Figure 1A is simple but only suitable for channels with high frequency selectivity or channels with high time fading. The method involves pilot extraction in the frequency domain (step 1A-1) followed by interpolation in time (step 1A-2), or interpolation in frequency (step 1A-3).

10 15 The second method shown in the flow chart of Figure 1B is aimed for channels with slow Doppler fading and fast frequency fading. It improves the first method by using FFT to reconstruct the channel response back to time domain for noise reduction processing at the expense of FFT/IFFT computing for the channel estimation separately. The method begins with pilot extraction in the frequency domain (step 1B-1), which may be followed by interpolation in frequency (step 1B-2). Then an inverse fast Fourier transform (step 1B-3), smoothing/de-noise processing (step 1B-4), and finally a 20 fast Fourier transform (1B-5) steps are executed.

25 The third method shown in the flow chart of Figure 1C can be used to estimate channel for mobile applications, where both fast time fading and frequency fading exist. However it needs a relatively high density of pilots and a completed interpolator. This method involves pilot extraction in the frequency domain (step 1C-1) this is followed by interpolation in time (step 1C-2) and interpolation in frequency (step 1C-3).

30 In the propagation environment with both high frequency dispersion and temporal fading, the channel estimation performance can be improved by the increase of pilot symbol density at the price of the reduction of the

spectral efficiency of the data transmission. To interpolate and reconstruct the channel response function from the limited pilots to achieve reliable channel estimation with the minimum overhead is a challenging task.

5 There are a variety of existing standard pilot patterns. In environments in which the channel varies only slowly with time and frequency, the pilot symbols may be inserted cyclically, being inserted at an adjacent frequency after each time interval. In environments in which the
10 channel is highly frequency dependent, the pilot symbols may be inserted periodically at all frequencies simultaneously. However, such a pilot pattern is only suitable for channels that vary very slowly with time. In environments in which the channel is highly time dependent, the pilot symbols may
15 be inserted continuously at only specific frequencies in a comb arrangement to provide a constant measurement of the channel response. However, such a pilot pattern is only suitable for channels that vary slowly with frequency. In environments in which the channel is both highly frequency
20 and highly time dependent (for example, mobile systems with much multi-path fading), the pilot symbols may be inserted periodically in time and in frequency so that the pilot symbols form a rectangular lattice when the symbols are depicted in a time-frequency diagram.

25 In OFDM communication systems employing coherent modulation and demodulation, the receiver must estimate the channel response at the frequencies of all sub-carriers and at all times. Although this requires more processing than in systems that employs differential modulation and
30 demodulation, a significant gain in signal-to-noise ratio can be achieved using coherent modulation and demodulation. The receiver determines the channel response at the times and frequencies at which pilot symbols are inserted into the

OFDM symbol, and performs interpolations to estimate the channel response at the times and frequencies at which the data symbols are located within the OFDM symbol. Placing pilot symbols more closely together (in frequency if a comb 5 pattern is used, in time if a periodic pattern is used, or in both frequency and in time if a rectangular lattice pattern is used) within a pilot pattern results in a more accurate interpolation. However, because pilot symbols are overhead, a tighter pilot pattern is at the expense of the 10 transmitted data rate.

Existing pilot patterns and interpolation techniques are usually sufficient if the channel varies slowly with time (for example for nomadic applications). However, if the channel varies quickly with time (for 15 example, for mobile applications), the time interval between pilot symbols must be reduced in order to allow an accurate estimation of the channel response through interpolation. This increases the overhead in the signal.

The problem of minimizing the number of pilot 20 symbols while maximizing the accuracy of the interpolation is also particularly cumbersome in Multiple-Input Multiple-Output (MIMO) OFDM systems. In MIMO OFDM systems, the transmitter transmits data through more than one transmitting antenna and the receiver receives data through 25 more than one receiving antenna. The binary data is usually divided between the transmitting antennae, although the same data may be transmitted through each transmitting antenna if spatial diversity is desired. Each receiving antenna receives data from all the transmitting antennae, so if 30 there are M transmitting antennae and N receiving antennae, then the signal will propagate over $M \times N$ channels, each of which has its own channel response. Each transmitting antenna inserts pilot symbols into the same sub-carrier

location of the OFDM symbol which it is transmitting. In order to minimize interference at the receiver between the pilot symbols of each transmitting antenna, each transmitting antenna typically blinks its pilot pattern on and off. This increases the temporal separation of the pilot symbols for each transmitter, reducing the accuracy of the interpolation used to estimate the channel response. In MIMO-OFDM systems a simple and fast channel estimation method is particularly crucial because of the limitation of the computational power for estimating $M \times N$ channels, while in SISO-OFDM system only one channel needs to be estimated.

SUMMARY OF THE INVENTION

Channel estimation methods are provided which are based on the partial interpolation of a scattered pilot by using true 2-D interpolation; and additionally, simple 1-D interpolation is used reconstruct the entire channels. This method has a reduced scattered pilot overhead, and is at least an order of magnitude less computationally complex than some existing methods. In general, the proposed method of channel estimation is more robust in channels with high Doppler spread, and provides better performance than some existing methods and requires the less buffering of the OFDM symbols for the coherent detection at the receiver than in some methods.

The methods allow fewer pilot symbols to be placed within each OFDM symbol, while still allowing accurate interpolation of the channel response. The data rate of an MIMO-OFDM system is thereby improved.

A first broad aspect of the invention provides a method of inserting pilot symbols into Orthogonal Frequency Division Multiplexing (OFDM) frames at an OFDM transmitter having at least one transmitting antenna, the OFDM frames

having a time domain and a frequency domain, each OFDM frame comprising a plurality of OFDM symbols. The method involves, for each antenna, inserting scattered pilot symbols in an identical scattered pattern in time-frequency.

5 In some embodiments, the identical scattered pattern is a regular diagonal-shaped lattice.

In some embodiments, inserting pilot symbols in an identical diagonal-shaped lattice involves for each point in the identical diagonal shaped lattice inserting a number of 10 pilot symbols on a single sub-carrier for N consecutive OFDM symbols, where N is the number of transmitting antennae.

In some embodiments, diagonal shaped lattice is a diamond shaped lattice.

In some embodiments for each point in the 15 diagonal-shaped lattice, L uncoded pilot symbols are generated. Space time block coding (STBC) is performed on the group of L uncoded pilot symbols to produce an NxN STBC block, L and N determining an STBC code rate. Then, one row or column of the STBC block is transmitted on each antenna 20 on a specific sub-carrier.

In some embodiments, transmitting the pilot symbols is done with a power level greater than a power level of data symbols, depending upon a value reflective of channel conditions.

25 In some embodiments, transmitting the pilot symbols is done with a power level which is dynamically adjusted to ensure sufficiently accurate reception as a function of a modulation type applied to the sub-carriers carrying data.

In some embodiments, the diagonal shaped lattice pattern has a first plurality of equally spaced sub-carrier positions, and a second plurality of equally spaced sub-carrier positions offset from said first plurality. The 5 pilot symbols are inserted alternately in time using the first plurality of equally spaced sub-carrier positions and the second plurality of equally spaced sub-carrier positions.

In some embodiments, the second plurality of sub-carriers is offset from the first plurality of equally spaced sub-carrier positions by half the spacing between adjacent sub-carriers of the first plurality of sub-carrier positions thereby forming a diamond shaped lattice pattern. 10

In some embodiments, the pilot pattern is 15 cyclically offset, both in a time direction and in a frequency direction, for at least one adjacent base station to form re-use patterns.

Another broad aspect of the invention provides an OFDM transmitter. The OFDM transmitter has a plurality of 20 transmit antennas, and is adapted to insert pilot symbols into Orthogonal Frequency Division Multiplexing (OFDM) frames having a time domain and a frequency domain, each OFDM frame comprising a plurality of OFDM symbols by, for each antenna, inserting pilot symbols in an identical 25 scattered pattern in time-frequency.

In some embodiments, the identical scattered pattern is a diagonal-shaped lattice.

In some embodiments, inserting pilot symbols in an identical scattered pattern involves for each point in the 30 identical scattered pattern inserting a number of pilot symbols on a single sub-carrier for N consecutive OFDM

symbols, where N is the number of transmitting antennae, where $N \geq 1$.

In some embodiments, the scattered pattern is a diamond shaped lattice.

5 In some embodiments, for each point in the scattered pattern, the OFDM transmitter is adapted to generate L uncoded pilot symbols, perform space time block coding (STBC) on the group of L pilot symbols to produce an NxN STBC block, and transmit one row or column of the STBC
10 block on each antenna.

In some embodiments, the OFDM transmitter is further adapted to transmit the pilot symbols with a power level greater than a power level of data symbols depending on a value reflective of channel conditions.

15 In some embodiments in which the diamond shaped lattice pattern is employed, the diamond shaped lattice pattern has a first plurality of equally spaced sub-carrier positions and a second plurality of equally spaced sub-carrier positions offset from said first plurality. The
20 pilot symbols are inserted alternately in time using the first plurality of equally spaced sub-carrier positions and the second plurality of equally spaced sub-carrier positions.

Another broad aspect of the invention provides a
25 method of estimating a plurality of channel responses at an Orthogonal Frequency Division Multiplexing (OFDM) receiver having at least one receive antenna. The method involves at each receive antenna receiving OFDM frames transmitted by at least one transmitting antenna, the OFDM frames having a
30 time domain and a frequency domain, the OFDM frames transmitted by each antenna having pilot symbols inserted in

an identical scattered pattern in time-frequency, each OFDM frame comprising a plurality of OFDM symbols. For each transmit antenna, receive antenna combination: a) the pilot symbols of the received OFDM frames are used to 5 estimate a channel response for each point in the scattered pattern; b) the channel response is estimated for a plurality of points not on the scattered pattern by performing a two-dimensional (time direction, frequency direction) interpolation of channel responses determined for 10 points in the scattered pattern; c) an interpolation is performed in the frequency direction to estimate the channel responses corresponding to remaining OFDM sub-carriers within each OFDM symbol.

In some embodiments, a filtering function is 15 performed on the channel responses prior to performing the interpolation in the frequency direction to estimate the channel responses corresponding to remaining OFDM sub-carriers within each OFDM symbol.

In some embodiments, estimating the channel 20 response of a plurality of points not on the scattered pattern by performing a two-dimensional (time direction, frequency direction) interpolation of channel responses determined for points in the scattered pattern lattice involves for each sub-carrier to be estimated averaging 25 channel responses of the given channel estimation period of a sub-carrier before the sub-carrier to be estimated in frequency (when present) and a sub-carrier after the sub-carrier to be estimated in frequency (when present) and the channel response for a previous estimation period (when 30 present) and a following estimation period (when present).

In some embodiments, the method is applied to a single transmitter, single receiver system.

In other embodiments the method is applied to a single transmitter system wherein each point in the scattered pattern contains a single pilot symbol.

5 In some embodiments, the method is applied to a system in which there are $N \geq 2$ antennas, and each point in the scattered pattern contains a number N of consecutive encoded pilot symbols transmitted on a sub-carrier, a single channel estimate being determined for each N encoded pilot symbols.

10 In some embodiments, the N encoded pilot symbols contain L pilot symbols which were STBC block coded, where N and L together determine a STBC code rate.

15 Other aspects and features of the present invention will become apparent to those ordinarily skilled in the art upon review of the following description of specific embodiments of the invention in conjunction with the accompanying Figures.

BRIEF DESCRIPTION OF THE DRAWINGS

20 The invention will now be described in greater detail with reference to the accompanying Figures, in which:

Figure 1 illustrates flow-charts for three examples of conventional OFDM Channel Estimation;

25 Figure 2 is a block diagram of a Multiple-Input Multiple-Output Orthogonal Frequency Division Multiplexing (OFDM) transmitter provided by an embodiment of the invention;

Figure 3 is a block diagram of an OFDM receiver;

Figure 4 is a flowchart of a method by which an OFDM transmitter inserts pilot symbols into an OFDM frame according to one embodiment of the invention;

Figure 5 is a diagram of a pilot pattern generated 5 using the method of Figure 4;

Figure 6 is a block diagram of a MIMO system showing the channel transfer functions between two transmit antennas and two receive antennas;

Figure 7 is a time frequency diagram showing 10 channel estimate positions for pilot channel estimation;

Figure 8 schematically illustrates a step of filtering estimated and interpolated pilot channel estimates;

Figure 9 shows schematically the step of 15 interpolating between the channel estimates previously determined to provide channel estimates for all sub-carriers and all times;

Figure 10 is a flow chart summarizing the overall 20 channel estimation method provided by an embodiment of the invention; and

Figure 11 is an example of a set of performance results obtained using the method of Figure 10.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

The following sections describe a MIMO-OFDM 25 transmitter/receiver and scattered pilot insertion. By way of introduction, a OFDM frame consists of the preamble OFDM symbols and regular OFDM symbols. Each OFDM symbol uses a set of orthogonal sub-carriers. When there are two transmit antennas, two OFDM symbols form a STTD block. For regular

OFDM symbols, some sub-carriers are used as pilot sub-carriers to carry pilot symbols while the others are used as data sub-carriers to carry data symbols. The pilot sub-carriers are modulated by pilot symbols generated by QPSK.

5 The data sub-carriers are modulated by complex data symbols generated by QAM mapping. STTD coding is applied to the pilot sub-carrier pairs located at the same frequency within one STTD block.

Referring to Figure 2, a block diagram of a

10 Multiple-Input Multiple-Output (MIMO) Orthogonal Frequency Division Multiplexing (OFDM) transmitter provided by an embodiment of the invention is shown. The OFDM transmitter shown in Figure 2 is a two-output OFDM transmitter, though more generally there may be a plurality of M transmitting

15 antennae. An OFDM transmitter 10 takes binary data as input but data in other forms may be accommodated. The binary data is passed to a coding/modulation primitive 12 responsible for encoding, interleaving, and modulating the binary data to generate data symbols, as is well known to

20 those skilled in the art. The coding/modulation primitive 12 may include a number of processing blocks, not shown in Figure 2. An encoder 14 applies Space-Time Block Coding (SBTC) to the data symbols. The encoder 14 also separates the data symbols into a first processing path 16 and a

25 second processing path 18, by sending alternate data symbols along each of the two processing paths. In the more general case in which the OFDM transmitter 10 includes M transmitting antennae, the encoder 14 separates the data symbols into M processing paths.

30 The data symbols sent along the first processing path 16 are sent to a first OFDM component 20. The data symbols are first passed to a demultiplexer 22 in the first OFDM component 20, after which the data symbols are treated

as sub-carrier components. The data symbols are then sent to a pilot inserter 24, where pilot symbols are inserted among the data symbols. Collectively, the data symbols and pilot symbols are referred to hereinafter simply as symbols.

5 The symbols are passed to an Inverse Fast Fourier Transform (IFFT) processor 26, then to a multiplexer 28 where they are recombined into a serial stream. A guard inserter 30 adds prefixes to the symbols. Finally, the OFDM signals are passed through a hard limiter 32, a digital-to-analog 10 converter 34, and a radio frequency (RF) transmitter 36 which transmits OFDM symbols as a signal through a first transmitting antenna 37. In most embodiments, each element in the first OFDM component 20 is a processor, a component of a larger processor, or a collection of processors or any 15 suitable combination of hardware, firmware and software. These might include general purpose processors, ASICs, FPGAs, DSPs to name a few examples.

The pilot inserter 24 is connected to receive space-time coded pilot symbols from pilot STBC function 23 20 which performs STBC on pilot symbols 21. The pilot STBC block 23 takes two pilot symbols at a time for example P_1 and P_2 as indicated in Figure 2 and generates an STBC block consisting of a two by two matrix having (P_1, P_2) in the first row and having $(-P_2^*, P_1^*)$ in the second row. It is 25 the first row of this STBC block that is inserted by the pilot inserter 24.

The data symbols sent along the second processing path 18 are sent to a second OFDM component 38 which includes processors similar to those included in the first 30 OFDM component 20. However, the pilot inserter 40 inserts encoded pilot symbols from the second row of the STBC block produced by the pilot STBC function 23. The symbols sent along the second processing path 18 are ultimately

transmitted as a signal through a second transmitting antenna 42.

Referring now to Figure 3, a block diagram of an MIMO-OFDM receiver is shown. An OFDM receiver 50 includes a 5 first receiving antenna 52 and a second receiving antenna 54 (although more generally there will be one or more receiving antennae). The first receiving antenna 52 receives a first received signal. The first received signal is a combination of the two signals transmitted by the two transmitting 10 antennae 37 and 42 of Figure 2, although each of the two signals will have been altered by a respective channel between the respective transmitting antenna and the first receiving antenna 52. The second receiving antenna 54 receives a second received signal. The second received 15 signal is a combination of the two signals transmitted by the two transmitting antennae 37 and 42 of Figure 2, although each of the two signals will have been altered by a respective channel between the respective transmitting antenna and the second receiving antenna 54. The four 20 channels (between each of the two transmitting antennae and each of the two receiving antennae) may vary with time and with frequency, and will usually be different from each other.

The OFDM receiver 50 includes a first OFDM 25 component 56 and a second OFDM component 58 (although in general there will be N OFDM components, one for each receiving antenna). The first OFDM component 56 includes a RF receiver 59, and an analog-to-digital converter 60, which converts the first received signal into digital signal 30 samples. The signal samples are passed to a frequency synchronizer 62 and a frequency offset corrector 64. The signal samples are also fed to a frame/time synchronizer 66.

Collectively, these three components produce synchronized signal samples.

The synchronized signal samples represent a time sequence of data. The synchronized signal samples are 5 passed to a demultiplexer 68, then passed in parallel to a Fast Fourier Transform (FFT) processor 70. The FFT processor 70 performs an FFT on the signal samples to generate estimated received symbols which are multiplexed in MUX 76 and sent as received symbols to decoder 78. Ideally, 10 the received symbols would be the same as the symbols fed into the IFFT processor 26 at the OFDM transmitter 10. However, as the received signals will have likely been altered by the various propagation channels, the first OFDM component 56 must correct the received symbols by taking 15 into account the channels. The received symbols are passed to a channel estimator 72, which analyses received pilot symbols located at known times and frequencies within the OFDM frame. The channel estimator 72 compares the received pilot symbols with what the channel estimator 72 knows to be 20 the values of the pilot symbols as transmitted by the OFDM transmitter 10, and generates an estimated channel response for each frequency and time within the OFDM symbol. The estimated channel responses are passed to decoder 78. The channel estimator 72 is described in detail below.

25 The second OFDM component 58 includes similar components as are included in the first OFDM component 56, and processes the second received signal in the same manner as the first OFDM component 56 processes the first received signal. Each OFDM component passes OFDM symbols to the 30 decoder 78.

The decoder 78 applies STBC decoding to the OFDM symbols, and passes the symbols to a decoding/demodulating

primitive 80 responsible for decoding, de-interleaving, and demodulating the symbols to generate output binary data, as is well known to those skilled in the art. The decoding/demodulation primitive 80 which may include a 5 number of additional processing blocks, not shown in FIG. 2. Each element in the OFDM components 56 and 58 is a processor, a component of a larger processor, or a collection of processors.

Referring now to Figure 4, a method by which each 10 of the pilot inserters 24 and 40 of Figure 2 inserts pilot symbols among the data symbols is shown. The method will be described with reference to the pilot inserter 24 in the first OFDM component 20. At step 100, the pilot inserter 24 receives data symbols from the demultiplexer 22. At step 15 102 the pilot STBC function 23 generates (or receives) two pilot symbols. At step 104 the pilot STBC function 23 applies STBC encoding to the pilot symbols, so as to generate an STBC block of encoded pilot symbols. The encoded pilot symbols generated for the first transmitting 20 antenna 37 will be one row of the STBC block and will have a number equal to the number of transmitting antennae in the OFDM transmitter. Thus, for a two antenna system a 2 x 2 25 STBC block is generated.

At step 106 the pilot inserter 24 inserts the 30 encoded pilot symbols within the OFDM symbol. Encoded pilot symbols are inserted in a diamond lattice pattern. The diamond lattice pattern uses the same frequencies as the other diamond lattice patterns, but has a temporal offset from the other diamond lattice patterns. Preferably, the temporal offset for each diamond lattice pattern is one symbol (in the time direction) from another diamond lattice pattern, so that the diamond lattice patterns use consecutive symbols in the time direction of the OFDM frame.

The diamond lattice pattern in which each encoded pilot symbol is inserted within the OFDM frame is preferably a perfect diamond lattice pattern. To achieve this, the encoded pilot symbol is inserted at each of a first subset 5 of frequencies. The frequencies within the first subset of frequencies are spaced equally apart by a pilot spacing. The encoded pilot symbol is inserted at each of the first subset of frequencies for an STBC block (two OFDM symbols). At some later time, the encoded pilot symbols are inserted 10 at each of a second subset of frequencies. The frequencies within the second subset of frequencies are shifted from the frequencies within the first subset of frequencies by half of the pilot spacing within the frequency direction. The pilot inserter 24 continues to insert encoded pilot symbols, 15 alternating between the first subset of frequencies and the second subset of frequencies.

Alternatively, a different pilot pattern can be used, as long as the same pilot pattern is used for each of the at least one encoded pilot symbols unique to the 20 transmitting antenna 37, and as long as the pilot patterns for the encoded pilot symbols are offset from each other in the time direction of the OFDM frame. For example, a regular diagonal lattice pattern may be used, the diamond shaped lattice being a special case of this.

25 The pilot inserter 40 inserts pilot symbols using the same method, although the pilot symbols will be the other half of the STBC block 42. The encoded pilot symbols unique to the second transmitting antenna 42 are inserted in the OFDM frame at the same symbol locations at which the 30 encoded pilot symbols corresponding to the first transmitting antenna 37 are inserted.

Referring to Figure 5, an example pilot pattern generated using the method of Figure 4 is shown. Pilot and data symbols are spread over the OFDM frame in a time direction 120 and a frequency direction 122. Most symbols 5 within the OFDM frame are data symbols 124. A first set of encoded pilot symbols 126 corresponding to the first transmitting antenna 37 is inserted in a diamond lattice pattern. A second set of encoded pilot symbols 128 corresponding to the first transmitting antenna 37 is 10 inserted in a diamond lattice structure at the same frequencies as the first set of encoded pilot symbols, but offset by one OFDM symbol location in the time direction 120. In the illustrated example two of every four OFDM symbols carry encoded pilot symbols. Each other 15 transmitting antenna transmits using the same pattern. The pairs of consecutive pilot symbols on a sub-carrier consist of two raw pilot symbols STBC encoded. The same pattern is transmitted by the second antenna.

The power of the encoded pilot symbols 126, 128 20 may be increased compared to the traffic data symbol 124. The power increase of the encoded pilot can be dynamically adjusted with respect to the transmitting data symbol power level or modulation type (QAM size), or as a function of channel quality. The location of diamond lattice pattern 25 may also be optimized to allow a fast extraction of scattered pilot without using the computing. This may be achieved if the pilot subcarriers are spaced in the frequency direction by 2^n . In the multiple base station transmission arrangement, the location of the diamond 30 lattice pattern can be cyclic offset both in time direction and in frequency direction amongst adjacent base stations to form a diamond lattice re-use pattern.

Referring now to Figures 6 to 10, a channel estimation method is described which is based on the pilot insertion method above. This invention presents a simple 2-dimensional channel interpolator for MIMO-OFDM system with 5 low pilot density for fast fading channels both in time and in frequency. The goal of channel estimation is to estimate the channel characteristics for each sub-carrier and at each time for each possible transmit antenna, receive antenna combination. Referring to Figure 13, for the two transmit 10 antenna, two receive antenna example, shown are two transmit antennas Tx1 140 and Tx2 142 and two receive antennas Rx1 144 and Rx2 146. Channel estimation estimates a channel for each sub-carrier and at each time between Tx1 140 and Rx1 144 indicated as each H_{11} 148, a channel between Tx1 140 and 15 Rx2 146 indicated by transfer function H_{12} 150, a channel estimate for transmitter Tx2 142 to Rx1 144 indicated as transfer function H_{22} 152 and finally, a channel estimate for transmitter Tx2 142 to receiver Rx2 146 indicated as transfer function H_{21} 154.

20 Some advantages for the proposed method compared to some existing methods are: (1) robust to high mobility-speed (2) a reduction of the scattered pilot grid density and therefore a reduction of the pilot overhead.

Let P_1 and P_2 be the two pilot symbols encoded in 25 an STBC block and transmitted by two antennas on one sub-carrier in consecutive OFDM symbols. Then at the first receive antenna, the following relationship exists for each sub-carrier on which pilot symbols are transmitted, where it is assumed the channel response H_{ij} is constant over two OFDM 30 frames:

$$\begin{bmatrix} Y_{1,1} \\ Y_{1,2} \end{bmatrix} = \begin{bmatrix} P_1 & P_2 \\ -P_2 & P_1 \end{bmatrix} \begin{bmatrix} H_{11} \\ H_{21} \end{bmatrix}$$

$Y_{1,1}$ is the received data on the first antenna on the sub-carrier in the first of the two consecutive OFDM symbols, and $Y_{1,2}$ is the received data on the first antenna on the sub-carrier in the second of the two consecutive symbols. This can be solved for H_{11} , H_{21} to yield:

$$\begin{bmatrix} H_{11} \\ H_{21} \end{bmatrix} = \frac{1}{|P_1|^2 + |P_2|^2} \begin{bmatrix} P_1^* & -P_2 \\ P_2^* & P_1 \end{bmatrix} \begin{bmatrix} Y_{1,1} \\ Y_{1,2} \end{bmatrix}$$

A similar process for the second antenna yields

$$\begin{bmatrix} H_{12} \\ H_{22} \end{bmatrix} = \frac{1}{|P_1|^2 + |P_2|^2} \begin{bmatrix} P_1^* & -P_2 \\ P_2^* & P_1 \end{bmatrix} \begin{bmatrix} Y_{2,1} \\ Y_{2,2} \end{bmatrix}$$

where $Y_{2,1}$ is the received data on the second antenna on the sub-carrier in the first of the two consecutive OFDM symbols, and $Y_{2,2}$ is the received data on the second antenna on the sub-carrier in the second of the two consecutive OFDM symbols.

Using this techniques, a channel estimate is made for each pilot sub-carrier, and for each pair of OFDM symbols used to transmit STBC blocks.

For the example of Figure 12, the result is a channel estimate, for each of the possible channels (these are for channels in this example as shown in Figure 13) for each pair of pilot symbols transmitted. This is illustrated in Figure 14 where only sub-carriers used to transmit pilots are shown. A channel estimate 150 is generated for each pair of (consecutive in time) OFDM frames for each pilot sub-carrier. This results in channel estimates 150, 152, 154 for the first and second frames, and channel estimates 156, 158, 160 for the fifth and sixth frames and so on.

The channel estimates are made on a STBC block by block basis so that the pattern of channel estimate shown in Figure 7 develops over time. The next step in the process is to perform an interpolation based on the channel estimate 5 of Figure 7 to obtain channel estimates for the places in Figure 7 which do not represent pilot channel positions. The manner in which this is done will be described for a single example, namely the unknown channel estimate indicated at 163 of Figure 7. Channel estimates are 10 buffered on an ongoing basis and when the four channel estimates 152, 156, 158 and 164 forming a diamond 162 surrounding the unknown channel estimate 163 have been computed, it is time to interpolate to obtain a channel estimate for the unknown point 163. The channel transfer 15 function at the sub-carrier located at the centre of the diamond can be obtained from a simple 4 points two-dimensional interpolator. Three points two-dimensional interpolators can be used to obtain the channel estimates corresponding to the first or last useful sub-carrier:

$$20 \quad H_{new}(n+1, k) = \frac{1}{4}(H(n, k) + H(n+2, k) + H(n+1, k-1) + H(n+1, k+1))$$

where ($k=2, \dots, N_{pilot}-1$)

$$H_{new}(n+1, 1) = \frac{1}{4}(H(n, 1) + H(n+2, 1) + 2H(n+1, 2))$$

$$H_{new}(n+1, N_{pilot}) = \frac{1}{4}(H(n, N_{pilot}) + H(n+2, N_{pilot}) + 2H(n, N_{pilot}-1))$$

where k is the pilot sub-carrier index, n is the channel 25 estimate index (or STBC block number - one channel estimate per sub-carrier for every two symbols), and N_{pilot} is the number of pilot sub-carriers (6 in the example of Figure 7). H_{new} is the newly interpolated channel estimate for the i^{th} channel estimation period, and the j^{th} pilot sub-carrier. 30 $H(i, j)$ is the channel estimate determined as described previously from the pilot symbols. A three points

interpolator would also be performed for the last STBC blocks in the OFDM frame (i.e. the last two OFDM symbols).

These calculations are done for each transmit antenna, receiver antenna combination. It is noted that 5 this is just one example of how the channel estimates can be interpolated.

If the original distance between pilot sub-carriers in the frequency direction is D_f , after first step of interpolation described above, the pilot sub-carriers' 10 separation becomes $\frac{D_f}{2}$.

In some embodiments, to remove noise, the channel estimates thus computed are filtered at each channel estimation period. This is shown in Figure 6 where the channel estimates 170 for one channel estimation period are 15 shown entering filter 172 to produce filtered channel estimates. For example, a simple 3 point moving iterative smoothing algorithm may be applied to H' :

$$H'_{sm}(n,k) = H'_{sm}(n,k-1) + \frac{1}{3}(H'(n,k+1) + H'_{sm}(n,k-2))$$

where $k=3, \dots, 2 N_{pilot}-2$. It is to be understood that other 20 filtering algorithms may be employed.

After the interpolation of the pilot channel estimate as summarized in Figure 7, there will be a channel estimate for each sub-carrier on which pilot channel information was transmitted and for each two OFDM symbol 25 period over which pilot channelling information was transmitted. Referring to Figure 5, this means that there will be a channel estimate for each antenna for time frequency points which are shaded to indicate that pilot channel information was transmitted. There will also be 30 channel estimates for the time frequency point in the centre

of the diamond shaped lattice structure of Figure 7. However, for points which are not pilot symbol transmission time-frequency points nor points which are at the centre of a diamond shaped lattice of such points, there will be no 5 channel estimate yet computed. The next step is to perform a further interpolation step to develop channel estimates for these other points.

In some embodiments, Cubic Lagrange interpolation and linear interpolation (for the sub-carriers near the 10 first and the last useful sub-carrier) in the frequency direction are used to obtain the channel transfer function at all sub-carriers for each STBC block (for each pair of OFDM symbols).

The coefficients of the Cubic Lagrange 15 interpolator can be calculated as

$$\begin{aligned}\mu(i) &= \frac{i}{D_f/2} & i = 1, 2, \dots, \frac{D_f}{2} \\ q_{-1}(\mu) &= -\frac{1}{6}\mu^3 + \frac{1}{2}\mu^2 - \frac{1}{3}\mu \\ q_0(\mu) &= \frac{1}{2}\mu^3 - \mu^2 - \frac{1}{2}\mu + 1 \\ q_1(\mu) &= -\frac{1}{2}\mu^3 + \frac{1}{2}\mu^2 + \mu \\ q_2(\mu) &= -\frac{1}{6}\mu^3 - \frac{1}{6}\mu\end{aligned}$$

The channel transfer functions at data sub- 20 carriers are given by

$$H_{\text{interp}}((j-1)\cdot\frac{D_f}{2} + i) = \sum_{n=-1}^2 q_n(\mu(i)) \cdot H_{\text{sm}}(j+n)$$

where $j=2, \dots, N_{\text{pilot}}-2$.

This is illustrated in Figure 9 where the 25 estimated channel responses are fed to the Lagrange cubic interpolator function 175 which outputs values for all intermediate sub-carriers. Other interpolations may alternatively be employed.

In some embodiments, every OFDM symbol contains some pilot insertion points and as such this completes the interpolation process. In other embodiments, there are some OFDM symbols which do not have any pilot insertion points.

5 To get channel estimates for these OFDM symbols, an interpolation in time of the previously computed channel estimates is performed. In high mobility applications, pilots should be included in every OFDM symbol avoiding the need for this last interpolation in time step.

10 Figure 10 presents an overall block diagram of the interpolation method proposed for two transmit antennas. An example set of performance results for the proposed MIMO-OFDM channel estimation algorithm is shown in Figure 10. The performance of the 2-D channel estimation algorithm is close 15 to the performance of ideal channel (only 0.5dB loss) at very high Doppler spread.

Referring now to Figures 10 and 3, the channel estimation method is carried out by the channel estimator 72 in order to estimate a channel response for each sub-carrier 20 and each OFDM symbol within an OFDM frame. The channel estimation method starts at step 500 by extracting the pilot symbols in the frequency domain for each receive antenna. This is followed by a channel response matrix computing step 502; whereby the received signal received by the receiving 25 antenna is decoded, which in effect performs a time average of the encoded pilot symbols at each point in the pilot pattern. For example, suppose the receiving antenna receives an OFDM frame having a pilot pattern as shown in Figure 5 (although the symbol 126 will now be a linear 30 combination of the encoded pilot symbol transmitted at this location by each of the transmitting antenna, and the symbol 128 will be a linear combination of the encoded pilot symbol transmitted at this location by each of the transmitting

antenna). Following decoding, the pilot symbol at symbol location 126 will be an average of the pilot symbol received at symbol location 126 and the pilot symbol received at symbol location 128. The time averaging effect produced by 5 the STBC decoding, during step 503, can be viewed as a pre-processing step, as can steps 500 and 502. The actual channel estimation method can be described broadly in four steps. Following step 503, during step 504 the channel estimator 72 estimates the channel response for each of a 10 plurality of pilot symbols. For a diamond lattice pattern, the plurality of pilot symbols will be four pilot symbols forming a single diamond pattern. The channel estimator 72 estimates the channel response of a central symbol, the central symbol having a time direction value and a frequency 15 direction value bounded by the time direction values and the frequency direction values of the plurality of pilot symbols. The central symbol preferably has a frequency direction value equal to the frequency direction values of two of the plurality of pilot symbols, and has a time 20 direction value midway between the time direction values of the two pilot symbols having the same frequency direction value as the central symbol. This can generally be described as a four-point 2-D interpolation of the channel response between pilot symbols. Third, the channel estimator 25 72 smoothes the channel responses (corresponding to both encoded pilot symbols and to the central symbol) in the frequency direction, preferably by performing a three-point smoothing, as per step 505. Fourth, the channel estimator 72 performs an interpolation in the frequency direction to 30 estimate the channel response for remaining symbols, as per step 506. The interpolation may be a linear interpolation for symbols having a frequency direction value equal to a first or a last useful sub-carrier within the OFDM symbol, and a cubic Lagrange interpolation otherwise.

The method of inserting pilot symbols (described above with reference to Figure 4) and the channel estimation method (described above with reference to Figure 10) need not be used together. Any channel estimation method may be 5 used by the OFDM receiver to estimate the channel responses for an OFDM frame containing encoded pilot symbols inserted using the method described above. However, due to the sparse distribution of the pilot symbols in the pilot pattern described above with reference to Figure 4 and 10 Figure 5, a two-dimensional interpolation method is preferable over a one-dimensional interpolation method. Similarly, the channel estimation method may be applied to an OFDM frame containing any pattern of pilot symbols.

The invention has been described with respect to 15 an MIMO-OFDM communication system. The invention may also be used with advantage in a single input-multiple output OFDM communication system, as the method of inserting pilot symbols (described with reference to Figure 4) and the channel estimation method (described with reference to 20 Figure 10) do not depend on the number of receiving antenna. Each receiving antenna within the OFDM receiver 50 performs channel estimation independently, regardless of the number of receiving antennae present.

The channel estimation method described with 25 reference to Figure 10 will also be advantageous in an OFDM communication system having only one transmitting antenna, as the method provides an improved interpolation of the channel response regardless of the number of transmitting antenna. The method of inserting pilot symbols described 30 with reference to Figure 11 may be used in an OFDM communication system having only one transmitting antenna, but will not be as advantageous as in an OFDM communication

system having more than one transmitting antenna as there will be no reduction in overhead.

The method of inserting pilot symbols and the channel estimation method are preferably implemented on the 5 OFDM transmitter and on the OFDM receiver respectively in the form of software instructions readable by a digital signal processor. Alternatively, the methods may be implemented as logic circuitry within an integrated circuit. More generally, any computing apparatus containing logic for 10 executing the described functionality may implement the methods. The computing apparatus which implements the methods (in particular the pilot inserter or the channel estimator) may be a single processor, more than one processor, or a component of a larger processor. The logic 15 may comprise external instructions stored on a computer-readable medium, or may comprise internal circuitry.

What has been described is merely illustrative of the application of the principles of the invention. Other arrangements and methods can be implemented by those skilled 20 in the art without departing from the spirit and scope of the present invention.

We Claim:

1. A method of inserting pilot symbols into Orthogonal Frequency Division Multiplexing (OFDM) frames at an OFDM transmitter having at least one transmitting antenna, the OFDM frames having a time domain and a frequency domain, each OFDM frame comprising a plurality of OFDM symbols, the method comprising the steps of:

for each antenna, inserting scattered pilot symbols in an identical scattered pattern in time-frequency.

10 2. A method according to claim 1 wherein the identical scattered pattern is a regular diagonal-shaped lattice.

15 3. A method according to claim 2 wherein for each antenna, inserting pilot symbols in an identical diagonal-shaped lattice comprises for each point in the identical diagonal shaped lattice inserting a number of pilot symbols on a single sub-carrier for N consecutive OFDM symbols, where N is the number of transmitting antennae.

20 4. A method according to claim 3 wherein the diagonal shaped lattice is a diamond shaped lattice.

5. A method according to claim 3 further comprising for each point in the diagonal-shaped lattice:

generating L uncoded pilot symbols;

25 performing space time block coding (STBC) on the group of L uncoded pilot symbols to produce an NxN STBC block, L and N determining an STBC code rate;

transmitting one row or column of the STBC block on each antenna on a specific sub-carrier.

6. A method of claim 1 further comprising transmitting the pilot symbols with a power level greater than a power level of data symbols, depending upon a value reflective of channel conditions.

5 7. A method of claim 4 further comprising transmitting the pilot symbols with a power level which is dynamically adjusted to ensure sufficiently accurate reception as a function of a modulation type applied to the sub-carriers carrying data.

10 8. A method according to claim 2 wherein the diagonal shaped lattice pattern comprises:

 a first plurality of equally spaced sub-carrier positions;

15 a second plurality of equally spaced sub-carrier positions offset from said first plurality;

 wherein the pilot symbols are inserted alternately in time using the first plurality of equally spaced sub-carrier positions and the second plurality of equally spaced sub-carrier positions.

20 9. A method according to claim 8 wherein the second plurality of sub-carriers is offset from the first plurality of equally spaced subcarrier positions by half the spacing between adjacent sub-carriers of the first plurality of sub-carrier positions thereby forming a diamond shaped lattice pattern.

25 10. The method of claim 1 wherein the pilot pattern is cyclically offset, both in a time direction and in a frequency direction, for at least one adjacent base station to form re-use patterns.

11. An OFDM transmitter comprising:

a plurality of transmit antennas;

the OFDM transmitter being adapted to insert pilot symbols into Orthogonal Frequency Division Multiplexing (OFDM) frames having a time domain and a frequency domain, each OFDM frame comprising a plurality of OFDM symbols by, for each antenna, inserting pilot symbols in an identical scattered pattern in time-frequency.

12. A transmitter according to claim 11 wherein the identical scattered pattern is a diagonal-shaped lattice.

13. An OFDM transmitter according to claim 11 wherein for each antenna, inserting pilot symbols in an identical scattered pattern comprises for each point in the identical scattered pattern inserting a number of pilot symbols on a single sub-carrier for N consecutive OFDM symbols, where N is the number of transmitting antennae, where $N \geq 1$.

14. An OFDM transmitter according to claim 13 wherein the scattered pattern is a diamond shaped lattice.

15. An OFDM transmitter to claim 11 further adapted to, for each point in the scattered pattern:

generate L uncoded pilot symbols;

perform space time block coding (STBC) on the group of L pilot symbols to produce an NxN STBC block;

transmit one row or column of the STBC block on each antenna.

16. An OFDM transmitter according to claim 15 wherein the scattered pattern is a diamond-shaped lattice.

17. An OFDM transmitter according to claim 11 further adapted to transmit the pilot symbols with a power level greater than a power level of data symbols depending on a value reflective of channel conditions.

5 18. An OFDM transmitter according to claim 11 further adapted to transmit the pilot symbols with a power level which is dynamically adjusted to ensure sufficiently accurate reception.

19. An OFDM transmitter according to claim 16 wherein
10 the diamond shaped lattice pattern comprises:

a first plurality of equally spaced sub-carrier positions;

a second plurality of equally spaced sub-carrier positions offset from said first plurality;

15 wherein the pilot symbols are inserted alternately in time using the first plurality of equally spaced sub-carrier positions and the second plurality of equally spaced sub-carrier positions.

20. An OFDM transmitter according to claim 19 wherein
20 the spacing between locations of the diamond lattice pattern are optimized to allow a fast extraction of scattered pilot symbols without requiring the computation of a complete FFT.

25. A method of estimating a plurality of channel responses at an Orthogonal Frequency Division Multiplexing (OFDM) receiver having at least one receive antenna, the method comprising:

at each receive antenna receiving OFDM frames transmitted by at least one transmitting antenna, the OFDM frames having a time domain and a frequency domain, the OFDM

frames transmitted by each antenna having pilot symbols inserted in an identical scattered pattern in time-frequency, each OFDM frame comprising a plurality of OFDM symbols;

5 for each transmit antenna, receive antenna combination:

a) using the pilot symbols of the received OFDM frames to estimate a channel response for each point in the scattered pattern;

10 b) estimating the channel response of a plurality of points not on the scattered pattern by performing a two-dimensional (time direction, frequency direction) interpolation of channel responses determined for points in the scattered pattern;

15 c) performing an interpolation in the frequency direction to estimate the channel responses corresponding to remaining OFDM sub-carriers within each OFDM symbol.

22. A method according to claim 21 further comprising:

20 performing a filtering function on the channel responses prior to performing the interpolation in the frequency direction to estimate the channel responses corresponding to remaining OFDM sub-carriers within each OFDM symbol

25 23. A method according to claim 21 wherein the scattered pattern is a regular diamond shaped lattice.

24. A method according to claim 23 wherein estimating the channel response of a plurality of points not on the scattered pattern by performing a two-dimensional (time direction, frequency direction) interpolation of channel

responses determined for points in the scattered pattern lattice comprises:

for each sub-carrier to be estimated averaging channel responses of the given channel estimation period of 5 a sub-carrier before the subcarrier to be estimated in frequency (when present) and a sub-carrier after the subcarrier to be estimated in frequency (when present) and the channel response for a previous estimation period (when present) and a following estimation period (when present).

10 25. A method of claim 22 wherein filtering the channel responses comprises performing a three-point smoothing operation.

26. A method of claim 21 wherein performing an interpolation in the frequency domain comprises performing a 15 linear interpolation for sub-carriers at a lowest or highest useful frequency within the OFDM symbol and performing a cubic Lagrange interpolation for sub-carriers at frequencies not equal to the first or the last useful frequency.

27. A method according to claim 21 applied to a single 20 transmitter, single receiver system.

28. A method according to claim 21 applied to a single transmitter system wherein each point in the scattered pattern contains a single pilot symbol.

29. A method according to claim 21 applied to a system 25 in which there are $N \geq 2$ transmit antennas, and wherein each point in the scattered pattern contains a number N of consecutive encoded pilot symbols transmitted on a sub-carrier, a single channel estimate being determined for each N encoded pilot symbols.

30. A method according to claim 29 wherein the N encoded pilot symbols contain L pilot symbols which were STBC block coded, where N and L together determine a STBC code rate.

5 31. A method according to claim 21 wherein the scattered pattern is a regular diagonal-shaped lattice.

32. A method according to claim 31 wherein the regular diagonal-shaped lattice is a diamond shaped lattice.

1/7

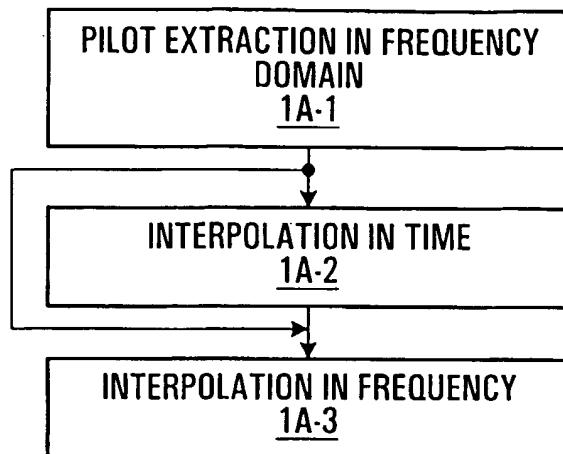


FIG. 1A
(PRIOR ART)

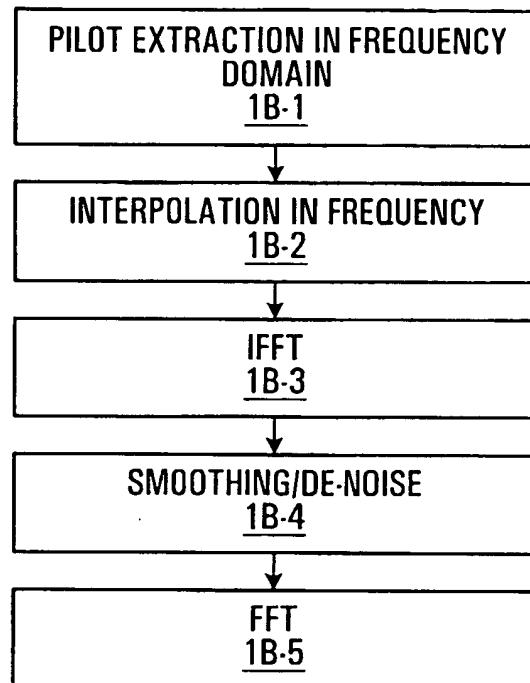


FIG. 1B
(PRIOR ART)

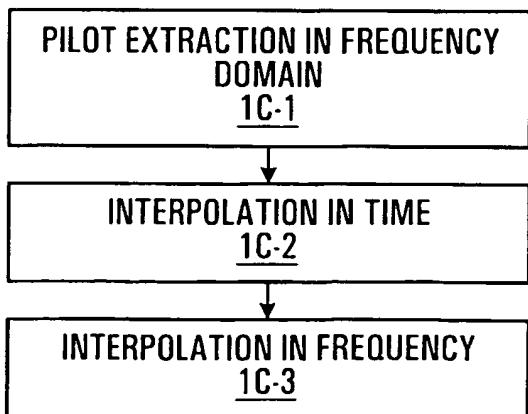


FIG. 1C
(PRIOR ART)

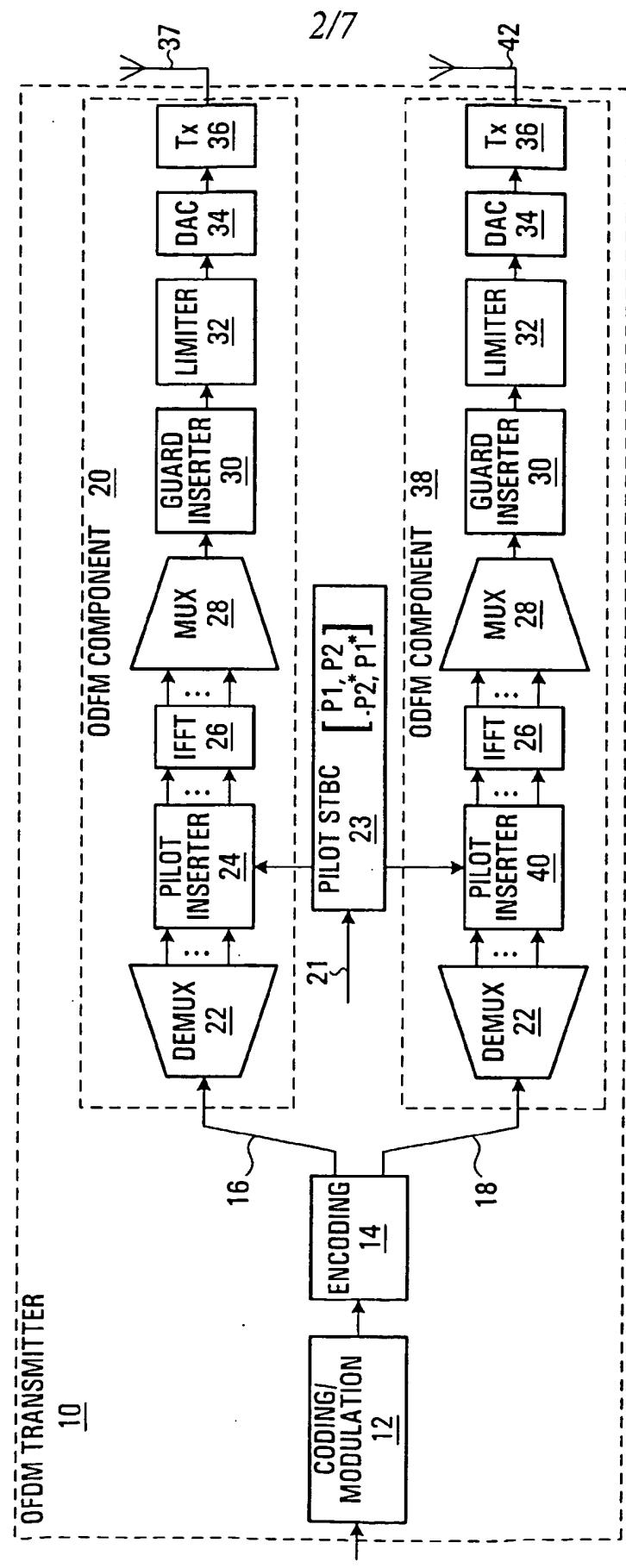


FIG. 2

3/7

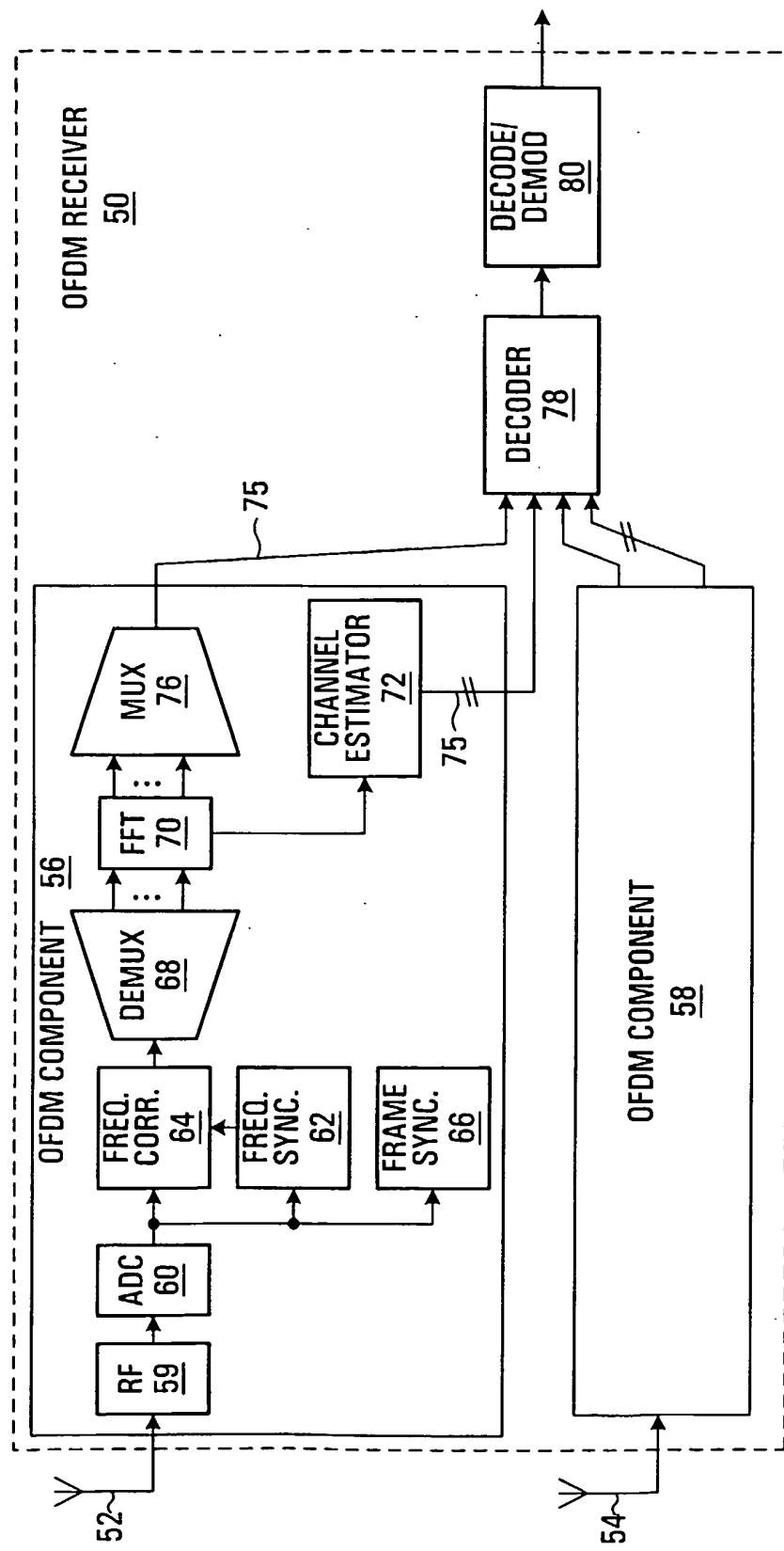


FIG. 3

4/7

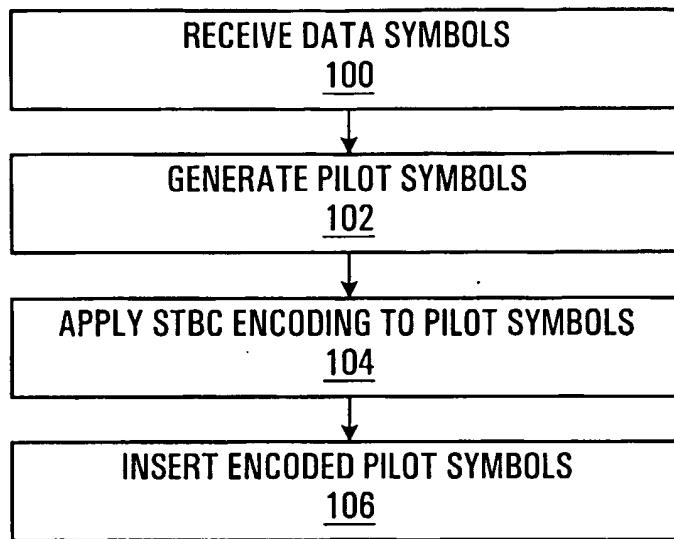


FIG. 4

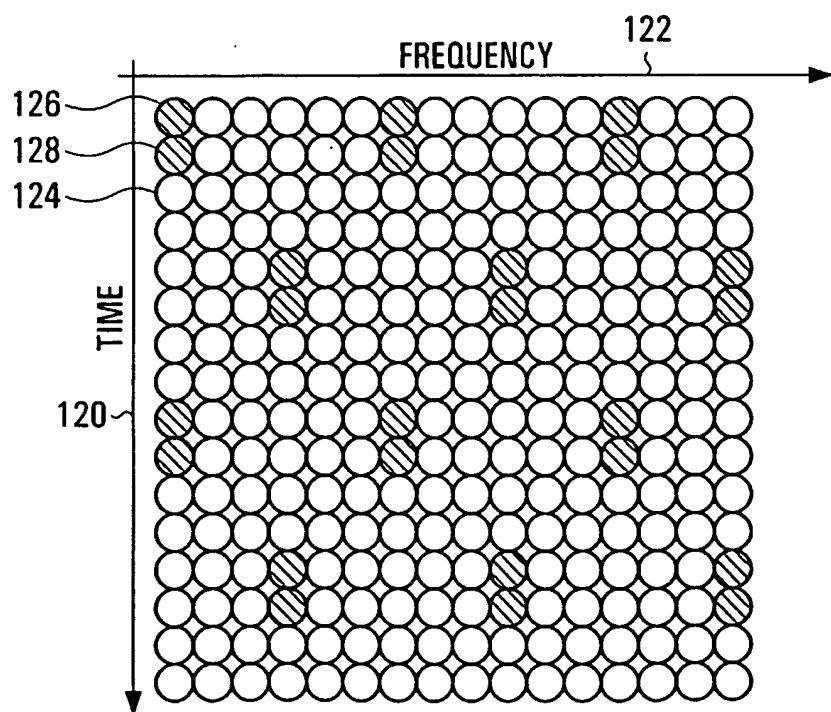


FIG. 5

5/7

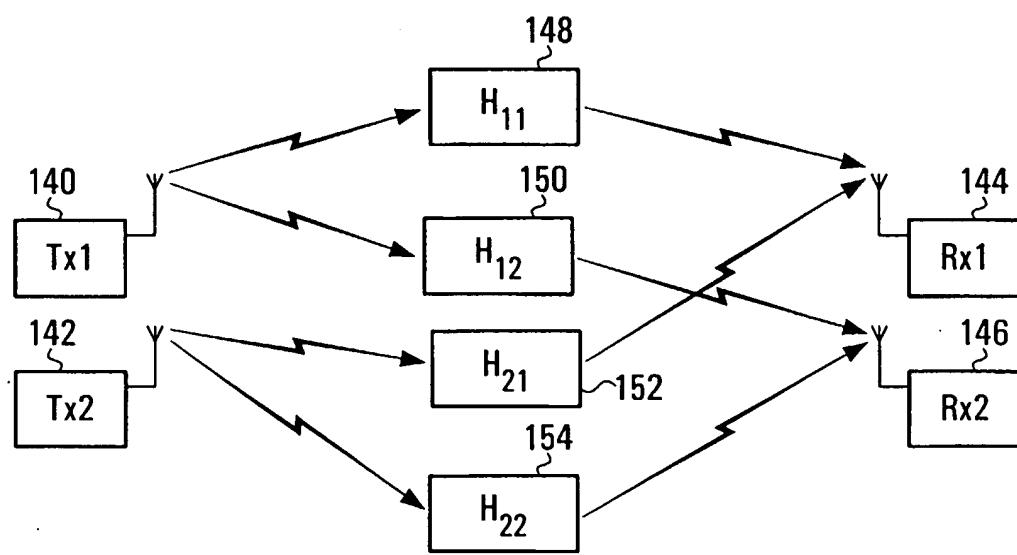
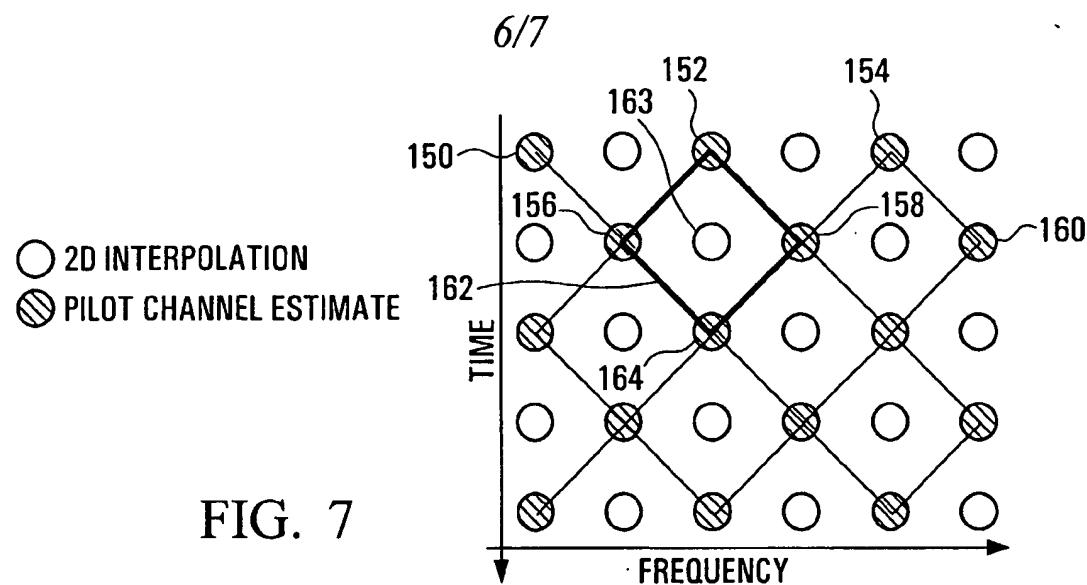


FIG. 6



7/7

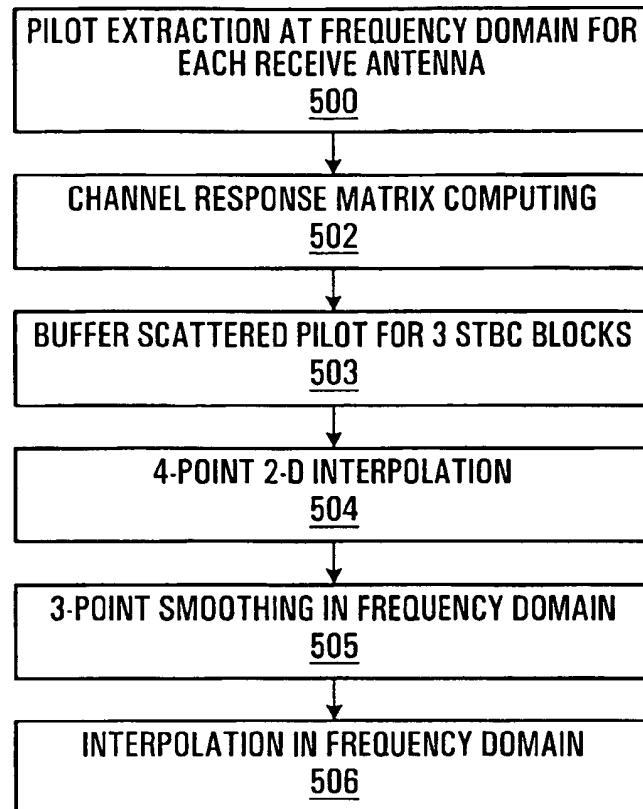


FIG. 10

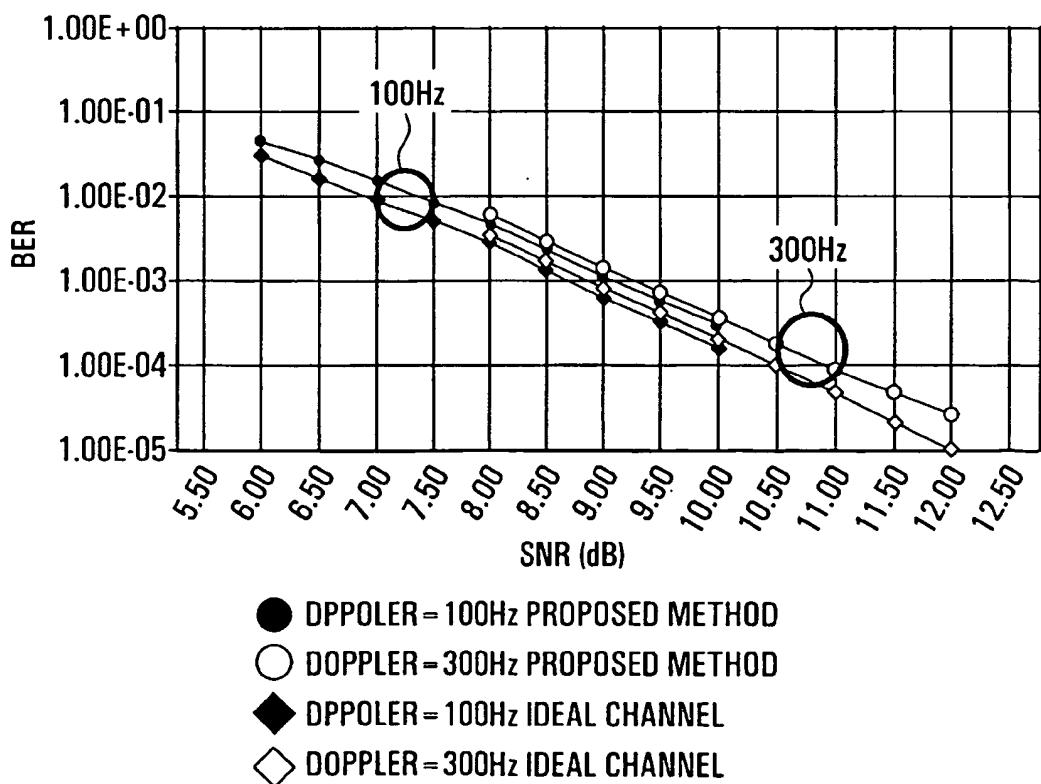


FIG. 11

INTERNATIONAL SEARCH REPORT

Int'l Application No

PCT/CA 02/01541

A. CLASSIFICATION OF SUBJECT MATTER

IPC 7 H04L5/02 H04L27/26 H04L25/02

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 7 H04L

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal, WPI Data, PAJ, INSPEC, COMPENDEX

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	US 6 298 035 B1 (HEISKALA JUHA) 2 October 2001 (2001-10-02) column 6, line 37 -column 8, line 33	1,11
Y		2-4,8,9, 12-14
A	figures 3,4,6 ---- -/-	5-7,10, 11,15-32

 Further documents are listed in the continuation of box C. Patent family members are listed in annex.

* Special categories of cited documents:

- *A* document defining the general state of the art which is not considered to be of particular relevance
- *E* earlier document but published on or after the international filing date
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- *O* document referring to an oral disclosure, use, exhibition or other means
- *P* document published prior to the international filing date but later than the priority date claimed

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- *X* document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone
- *Y* document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.
- *&* document member of the same patent family

Date of the actual completion of the international search

20 January 2003

Date of mailing of the international search report

30/01/2003

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INTERNATIONAL SEARCH REPORT

Int.....ional Application No

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C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
Y	FERNANDEZ-GETINO GARCIA J ET AL: "EFFICIENT PILOT PATTERNS FOR CHANNEL ESTIMATION IN OFDM SYSTEMS OVER HF CHANNELS" VTC 1999: GATEWAY TO THE 21ST. CENTURY COMMUNICATIONS VILLAGE, vol. 4 CONF. 50, 19 - 22 September 1999, pages 2193-2197, XP000896002 NEW YORK, NY, USA ISBN: 0-7803-5436-2 A page 2194, left-hand column, line 4 -right-hand column, line 8 figure 2 ---	2-4,8,9, 12-14
A	JONES V K ET AL: "CHANNEL ESTIMATION FOR WIRELESS OFDM SYSTEMS" IEEE GLOBECOM 1998, vol. 2, 8 - 12 November 1998, pages 980-985, XP000825895 NEW YORK, NY, USA ISBN: 0-7803-4985-7 abstract page 980, right-hand column, line 29 -page 981, right-hand column, line 6 page 981, right-hand column, line 29 -page 982, left-hand column, line 10 page 982, left-hand column, line 38 - line 45 page 983, right-hand column, line 9 - line 33 ----	1,11,21
A		1,11,21

INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No

PCT/CA 02/01541

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